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## Model reduction for controller design for infinite-dimensional systems

Opmeer, Mark Robertus

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*Document Version*

Publisher's PDF, also known as Version of record

*Publication date:*  
2006

[Link to publication in University of Groningen/UMCG research database](#)

*Citation for published version (APA):*

Opmeer, M. R. (2006). *Model reduction for controller design for infinite-dimensional systems*. [Thesis fully internal (DIV), University of Groningen]. s.n.

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# Chapter 8

## Robust stabilization

In this chapter we consider so-called frequency domain feedback controller design. We are mainly interested in feedback controllers that provide a certain type of robustness. The following definition is basic for this chapter.

**Definition 8.1.** Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . We say that  $K$  is an **admissible feedback function** for  $G$  if  $K : D(K) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{Y}, \mathcal{U})$  is holomorphic with  $0 \in D(K)$  and  $I - KG$  has a bounded inverse in zero.

**Lemma 8.2.** *Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$  and let  $K$  be an admissible feedback function for  $G$ . Then  $I - GK$  has a bounded inverse in zero.*

*Proof.* This follows from Lemma 3.16 with  $Z = G(0)$ ,  $T = K(0)$  and  $\lambda = 1$ .  $\square$

*Remark 8.3.* Note that it follows from Lemma 8.2 that  $K$  is an admissible feedback function for  $G$  if and only if  $G$  is an admissible feedback function for  $K$ .

**Definition 8.4.** An admissible feedback function  $K$  for  $G$  is called **stabilizing** if

$$\begin{bmatrix} (I - KG)^{-1} & K(I - GK)^{-1} \\ G(I - KG)^{-1} & (I - GK)^{-1} \end{bmatrix} \in H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{U} \times \mathcal{Y}, \mathcal{U} \times \mathcal{Y})). \quad (8.1)$$

*Remark 8.5.* Note that the function in (8.1) is the inverse of

$$\begin{bmatrix} I & -K \\ -G & I \end{bmatrix}.$$

The intuition behind the above definitions is that  $G$  is the transfer function of the plant and that  $K$  is the transfer function of the controller. The interconnection shown in figure 8.1 of the two systems is well-defined when  $K$  is an admissible feedback function for  $G$ . The condition (8.1) is equivalent to the transfer function from  $[u_1; u_2]$  to  $[e_1; e_2]$  being in  $H^\infty$ .

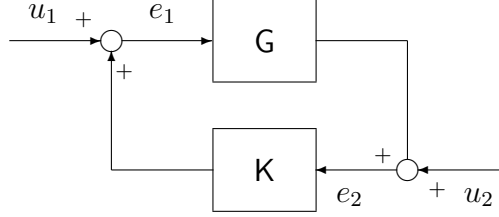


Figure 8.1: Feedback interconnection of  $G$  and  $K$ .

The following proposition, and most of the other propositions in this chapter, are formulated for right factorizations. By applying them to  $G^\dagger$  one obtains the obvious left versions of the results.

**Proposition 8.6.** *Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . Assume that  $G$  has a strongly right-coprime factor  $[M; N]$ . Let  $[\tilde{X}, \tilde{Y}]$  be a right Bezout pair for this factorization and assume that  $\tilde{X}$  has a bounded inverse in zero. Define  $K := \tilde{X}^{-1}\tilde{Y}$ . Then  $K$  is a stabilizing admissible feedback function for  $G$ .*

*Proof.* We have

$$I - KG = I - \tilde{X}^{-1}\tilde{Y}NM^{-1} = \tilde{X}^{-1}(\tilde{X}M - \tilde{Y}N)M^{-1} = \tilde{X}^{-1}M^{-1},$$

which shows that  $I - KG$  has a bounded inverse in zero. Furthermore, we have

$$\begin{bmatrix} (I - KG)^{-1} & K(I - GK)^{-1} \\ G(I - KG)^{-1} & (I - GK)^{-1} \end{bmatrix} = \begin{bmatrix} M\tilde{X} & M\tilde{Y} \\ N\tilde{X} & I + N\tilde{Y} \end{bmatrix},$$

which is in  $H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{U} \times \mathcal{Y}, \mathcal{U} \times \mathcal{Y}))$ . To obtain the formula for  $(I - GK)^{-1}$  we have used that

$$(I - GK)^{-1} = I + G(I - KG)^{-1}K,$$

which is easily checked. □

**Proposition 8.7.** *Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . Assume that there exists a stabilizing admissible feedback function  $K$  for  $G$ . Then  $G$  has a strongly right-coprime factorization.*

*Proof.* Define  $M_1 := (I - KG)^{-1}$  and  $N_1 := G(I - KG)^{-1}$ . Since the feedback function is stabilizing we have  $M_1, N_1 \in H^\infty$  and obviously  $G = N_1 M_1^{-1}$ . So  $G$  has a right factorization. By Proposition 7.13  $G$  has a weakly right-coprime factorization  $[M; N]$ . Similarly  $K$  has a weakly right-coprime factorization:  $K = YX^{-1}$ . Clearly,

$$N_2 := \begin{bmatrix} 0 & Y \\ N & 0 \end{bmatrix}, \quad M_2 := \begin{bmatrix} M & 0 \\ 0 & X \end{bmatrix}$$

provides a weakly right-coprime factorization of  $G_2 := [0, K; G, 0]$ . Since the feedback function is stabilizing we have  $(I - G_2)^{-1} \in H^\infty$ . It follows from Lemma 7.8 that  $(M_2 - N_2)^{-1} \in H^\infty$ . Denote the upper row of  $(M_2 - N_2)^{-1}$  by  $[\tilde{X}, \tilde{Y}]$ , then  $\tilde{X}M - \tilde{Y}N = I$ . Hence  $[M; N]$  is a strongly right-coprime factor.  $\square$

**Corollary 8.8.** *Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . Any stabilizing feedback function for  $G$  has a strongly right-coprime factorization.*

*Proof.* The symmetry mentioned in remark 8.3 and Proposition 8.7 give the result.  $\square$

**Proposition 8.9.** *Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . Then the following are equivalent:*

1.  $G$  has a strongly right-coprime factorization.
2. There exists a stabilizing admissible feedback function for  $G$ .

*Proof.* (2) implies (1) is Proposition 8.7. (1) implies (2) follows from Proposition 8.6 using Lemma 7.23.  $\square$

**Proposition 8.10.** *Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . Assume that  $G$  has a doubly coprime factorization. Then all stabilizing admissible feedback functions are given by*

$$K = (Y + MV)(X + NV)^{-1}, \quad (8.2)$$

where  $V \in H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{Y}, \mathcal{U}))$  is such that  $X + NV$  has a bounded inverse in zero, but is otherwise arbitrary.

*Proof.* We first show that the function defined by (8.2) is a stabilizing admissible feedback function. Define  $\underline{Y} := Y + MV$ ,  $\underline{X} := X + NV$ . Then

$$\begin{bmatrix} M & \underline{Y} \\ N & \underline{X} \end{bmatrix} = \begin{bmatrix} M & Y \\ N & X \end{bmatrix} \begin{bmatrix} I & V \\ 0 & I \end{bmatrix},$$

from which it easily follows that with  $\tilde{\underline{X}} := \tilde{X} + V\tilde{N}$  and  $\tilde{\underline{Y}} := \tilde{Y} + V\tilde{M}$  we obtain (7.3) with the Bezout factors replaced by the underlined versions. It follows from the left version of Proposition 8.6 that if  $\underline{X}$  is invertible at zero, then  $\underline{Y}\underline{X}^{-1}$  is a stabilizing admissible feedback function.

Assume that  $K$  is a stabilizing admissible feedback function. We will show that it is of the form (8.2). Since there exists a stabilizing admissible feedback function,  $G$  has a doubly coprime factorization by Propositions 7.24 and 8.7. From Proposition 7.24 and Corollary 8.8 we obtain that  $K$  has a strongly left-coprime factorization:  $K = \tilde{W}^{-1}\tilde{Z}$  and  $\tilde{W}S - \tilde{Z}R = I$ . Define  $\Delta := \tilde{Z}N - \tilde{W}M$ . Since  $\Delta = \tilde{W}(KG - I)M$  we have that  $\Delta$  is invertible in zero. It is easily calculated that the matrix in (8.1) can be written as

$$\begin{bmatrix} (I - KG)^{-1} & K(I - GK)^{-1} \\ G(I - KG)^{-1} & (I - GK)^{-1} \end{bmatrix} = \begin{bmatrix} -M\Delta^{-1}\tilde{W} & -M\Delta^{-1}\tilde{Z} \\ -N\Delta^{-1}\tilde{W} & I - N\Delta^{-1}\tilde{Z} \end{bmatrix}. \quad (8.3)$$

Using the above Bezout equation for the strongly left-coprime factorization of  $K$  we obtain  $M\Delta^{-1} = M\Delta^{-1}\tilde{W}S - M\Delta^{-1}\tilde{Z}R$ , which is in  $H^\infty$ . Similarly we obtain  $N\Delta^{-1} \in H^\infty$ . Using Proposition 7.7 we see that  $\Delta^{-1} \in H^\infty$ . Define  $V := \Delta^{-1}(\tilde{W}Y - \tilde{Z}X) \in H^\infty$ . We show that  $K(X + NV) = Y + MV$ . Using (8.3) we obtain

$$\begin{aligned} Y + MV &= Y + M\Delta^{-1}(\tilde{W}Y - \tilde{Z}X) = (I + M\Delta^{-1}\tilde{W})Y - M\Delta^{-1}\tilde{Z}X \\ &= (I - (I - KG)^{-1})Y + K(I - GK)^{-1}X, \end{aligned}$$

and

$$\begin{aligned} X + NV &= X + N\Delta^{-1}(\tilde{W}Y - \tilde{Z}X) = N\Delta^{-1}\tilde{W}Y + (I - N\Delta^{-1}\tilde{Z})X \\ &= -G(I - KG)^{-1}Y + (I - GK)^{-1}X. \end{aligned} \quad (8.4)$$

From this it easily follows that  $K(X + NV) = Y + MV$ . So the only thing left to show is that  $X + NV$  has a bounded inverse in zero. Using (8.4) we obtain

$$\tilde{M}(I - GK)(X + NV) = -\tilde{N}Y + \tilde{M}X = I,$$

where the last identity follows from (7.3). Since  $\tilde{M}(I - GK)$  has a bounded inverse in zero it follows that  $X + NV$  does and that this inverse equals  $\tilde{M}(I - GK)$ .  $\square$

**Lemma 8.11.** *If in a Banach algebra we have  $\|x\|^2 + \|y\|^2 \leq \alpha^2 < 1$ , then  $I - y$  is invertible and  $\|(I - y)^{-1}x\|^2 \leq \alpha^2/(1 - \alpha^2)$ .*

*Proof.* That  $I - y$  has a bounded inverse follows from the Neumann series theorem. From this theorem we also obtain  $\|(I - y)^{-1}\| \leq 1/(1 - \|y\|)$ . It follows that  $\|(I - y)^{-1}x\|^2 \leq \|x\|^2/(1 - \|y\|)^2$ . Denote  $x_1 := \|x\|$  and  $y_1 := \|y\|$ . Using elementary vector calculus one sees that the function  $x_1^2/(1 - y_1)^2$  under the constraint  $x_1^2 + y_1^2 \leq \alpha^2 < 1$  has the maximum  $\alpha^2/(1 - \alpha^2)$ . The desired result follows.  $\square$

We now focus our attention on feedback functions that provide a certain robustness.

**Definition 8.12.** Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . Suppose that  $G$  has a normalized weakly right-coprime factor  $[M; N]$ . Let  $\varepsilon > 0$ . The function  $G_\Delta$  is called a  **$\varepsilon$ -right factor admissible perturbation** of  $G$  if  $G_\Delta$  has a right factor  $[M_\Delta; N_\Delta]$  with  $\|[M_\Delta; N_\Delta] - [M; N]\|_\infty < \varepsilon$ .

Note that, due to Proposition 7.15, the set of  $\varepsilon$ -right factor admissible perturbations does not depend on the specific normalized weakly right-coprime factor chosen to define it.

**Definition 8.13.** Suppose that  $G$  has a normalized weakly right-coprime factor  $[M; N]$ . Let  $\varepsilon > 0$ . An admissible feedback function  $K$  for  $G$  is called  **$\varepsilon$ -robust right factor stabilizing** if it is a stabilizing admissible feedback function for all  $\varepsilon$ -right factor admissible perturbations of  $G$ .

**Proposition 8.14.** *Suppose that  $G$  has a normalized doubly coprime factorization. Let  $\varepsilon \in (0, 1)$ . Suppose that  $[\tilde{V}, \tilde{U}] \in H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{U} \times \mathcal{Y}, \mathcal{U}))$  satisfies*

$$\left\| [M^*, N^*] - [\tilde{V}, -\tilde{U}] \right\| \leq \sqrt{1 - \varepsilon^2},$$

*and that  $\tilde{V}$  has a bounded inverse in zero. Then  $K := \tilde{V}^{-1}\tilde{U}$  is a  $\varepsilon$ -robust right factor stabilizing admissible feedback function for  $G$ .*

*Proof.* Let  $W : \mathbb{T} \rightarrow \mathcal{L}(\mathcal{U} \times \mathcal{Y})$  be the function from Proposition 7.33, i.e.

$$W(z) = \begin{bmatrix} M(z) & -\tilde{N}(z)^* \\ N(z) & \tilde{M}(z)^* \end{bmatrix}.$$

Define  $F \in L^\infty(\mathbb{T}, \mathcal{L}(\mathcal{U} \times \mathcal{Y}, \mathcal{U}))$  by

$$F := \left( [M^*, N^*] - [\tilde{V}, -\tilde{U}] \right) W = [I - \tilde{V}M + \tilde{U}N, \tilde{V}\tilde{N}^* + \tilde{U}\tilde{M}^*], \quad (8.5)$$

where we have used (7.3). Since  $W(z)$  is unitary we have

$$\|F\|_\infty \leq \sqrt{1 - \varepsilon^2}.$$

It follows that  $\|I - \tilde{V}M + \tilde{U}N\|_\infty < 1$ . Since  $H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{U}))$  is a Banach algebra it follows that  $\tilde{V}M - \tilde{U}N$  has an inverse in  $H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{U}))$ . Denote an arbitrary  $\varepsilon$ -right factor admissible perturbation of  $G$  by  $G_\Delta$ . Denote a right factor of  $G_\Delta$  as in Definition 8.12 by  $[M_\Delta; N_\Delta]$ . Further denote  $\Delta = [\Delta_M; \Delta_N] = [M_\Delta; N_\Delta] - [M; N]$ . We have

$$\begin{aligned} I - KG_\Delta &= I - \tilde{V}^{-1}\tilde{U}(N + \Delta_N)(M + \Delta_M)^{-1} \\ &= \tilde{V}^{-1} \left( \tilde{V}M + \tilde{V}\Delta_M - \tilde{U}N - \tilde{U}\Delta_N \right) (M + \Delta_M)^{-1} \\ &= \tilde{V}^{-1} \left( \tilde{V}M - \tilde{U}N + [\tilde{V}, -\tilde{U}] \begin{bmatrix} \Delta_M \\ \Delta_N \end{bmatrix} \right) (M + \Delta_M)^{-1} \\ &= \tilde{V}^{-1}(\tilde{V}M - \tilde{U}N) (I + S\Delta) (M + \Delta_M)^{-1}, \end{aligned}$$

where

$$S := (\tilde{V}M - \tilde{U}N)^{-1}[\tilde{V}, -\tilde{U}].$$

It follows that  $I - KG_\Delta$  has an inverse in  $H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{U}))$  if and only if  $I + S\Delta$  does. The latter is true if  $\|S\|_\infty < 1/\varepsilon$ . We have

$$\begin{aligned} \|S\|_\infty^2 &= \|SW\|_\infty^2 = \|[I, -(\tilde{V}M - \tilde{U}N)^{-1}(\tilde{V}\tilde{N}^* + \tilde{U}\tilde{M}^*)]\|^2 \\ &= 1 + \|(\tilde{V}M - \tilde{U}N)^{-1}(\tilde{V}\tilde{N}^* + \tilde{U}\tilde{M}^*)\|^2 = 1 + \|(I - F_1)^{-1}F_2\|^2, \end{aligned}$$

where  $F = [F_1, F_2]$  is the function from (8.5). From Lemma 8.11 we obtain

$$\|S\|_\infty^2 \leq \frac{1}{\varepsilon^2},$$

as desired.

To check that  $K$  stabilizes  $G_\Delta$  we have to show that

$$\begin{bmatrix} (I - KG_\Delta)^{-1} & K(I - G_\Delta K)^{-1} \\ G_\Delta(I - KG_\Delta)^{-1} & (I - G_\Delta K)^{-1} \end{bmatrix} \in H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{U} \times \mathcal{U})).$$

We already saw that  $(I - KG_\Delta)^{-1} \in H^\infty$ . We compute

$$G_\Delta(I - KG_\Delta)^{-1} = (N + \Delta_N)(I + S\Delta)^{-1}(\tilde{V}M - \tilde{U}N)^{-1}\tilde{V},$$

which is clearly in  $H^\infty$ . Similarly we have

$$K(I - G_\Delta K)^{-1} = (I - KG_\Delta)^{-1}K = (M + \Delta_M)(I + S\Delta)^{-1}(\tilde{V}M - \tilde{U}N)^{-1}\tilde{U},$$

and

$$(I - G_\Delta K)^{-1} = I + G_\Delta(I - KG_\Delta)^{-1}K = I + (N + \Delta_N)(I + S\Delta)^{-1}(\tilde{V}M - \tilde{U}N)^{-1}\tilde{U},$$

which are both in  $H^\infty$ .  $\square$

**Proposition 8.15.** *Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . Suppose that  $G$  has a normalized doubly coprime factorization. Let  $H_{[M;N]}$  denote the Hankel operator of the normalized strongly right-coprime factor  $[M; N]$ . Let  $\varepsilon < \sqrt{1 - \|H_{[M;N]}\|^2}$ . Then there exists a  $[\tilde{V}, \tilde{U}] \in H^\infty(\mathbb{D}, \mathcal{L}(\mathcal{U} \times \mathcal{Y}, \mathcal{U}))$  such that*

$$\left\| [M^*, N^*] - [\tilde{V}, -\tilde{U}] \right\| \leq \sqrt{1 - \varepsilon^2}.$$

*If the input space  $\mathcal{U}$  is finite-dimensional, then  $[\tilde{V}, \tilde{U}]$  can be chosen such that  $\tilde{V}$  has a bounded inverse in zero.*

*Proof.* The existence of  $[\tilde{V}, \tilde{U}]$  such that the desired inequality is satisfied follows easily from Nehari's theorem (Lemma A.27). We now show that if  $\mathcal{U}$  is finite-dimensional we can choose  $\tilde{V}$  such that it has a bounded inverse in zero. First choose a  $\tilde{\varepsilon} \in (\varepsilon, \sqrt{1 - \|H_{[M;N]}\|^2})$  and find a  $[\tilde{V}_1, \tilde{U}_1]$  such that the desired inequality is satisfied with  $\tilde{\varepsilon}$  instead of  $\varepsilon$ . Define  $[\tilde{V}_\delta, \tilde{U}_\delta] := [\tilde{V}_1, \tilde{U}_1] - \delta[I, 0]$ . Then this satisfies

$$\left\| [M^*, N^*] - [\tilde{V}_\delta, -\tilde{U}_\delta] \right\| \leq \sqrt{1 - \tilde{\varepsilon}^2} + \delta.$$

It follows that for  $\delta \in (0, \sqrt{1 - \varepsilon^2} - \sqrt{1 - \tilde{\varepsilon}^2})$  the desired inequality is satisfied. Since  $\mathcal{U}$  is finite-dimensional, this interval must contain a point in the resolvent set of  $\tilde{V}_1(0)$ . For such a  $\delta$  we have that  $\tilde{V}_\delta$  has a bounded inverse in zero.  $\square$

*Remark 8.16.* If  $G(0) = 0$ , then the conclusion of Proposition 8.15 is also true without the assumption that the input space  $\mathcal{U}$  is finite-dimensional. We indicate why this is true. From the proof of Proposition 8.14 we obtain that  $\tilde{V}M - \tilde{U}N$  has an inverse in  $H^\infty$ . It follows that  $(\tilde{V}M - \tilde{U}N)(0)$  has a bounded inverse. We have  $N(0) = G(0)M(0) = 0$  and so  $\tilde{V}(0)M(0) = (\tilde{V}M - \tilde{U}N)(0)$  has a bounded inverse. Since  $M(0)$  has a bounded inverse it follows that  $\tilde{V}$  has a bounded inverse in zero as desired.

**Proposition 8.17.** *Let  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  be holomorphic with  $0 \in D(G)$ . Suppose that  $G$  has a normalized doubly coprime factorization and that  $\mathcal{U}$  is finite-dimensional. Let  $H_{[M;N]}$  denote the Hankel operator of the normalized strongly right-coprime factor  $[M; N]$ . Let  $\varepsilon \in (0, \sqrt{1 - \|H_{[M;N]}\|^2})$ . Then there exists an  $\varepsilon$ -robust right factor stabilizing admissible feedback function for  $G$ .*

*Proof.* This follows from combining Propositions 8.14 and 8.15.  $\square$



*Remark 8.18.* The assumption in Proposition 8.17 that  $\mathcal{U}$  is finite-dimensional is made because that assumption had to be made in Proposition 8.15 to obtain invertibility of  $\tilde{V}$  in zero. We can avoid making this assumption in Proposition 8.17 by considering controllers with internal loop as in Curtain, Weiss and Weiss [17].

**Corollary 8.19.** *Suppose that  $G : D(G) \subset \mathbb{C} \rightarrow \mathcal{L}(\mathcal{U}, \mathcal{Y})$  with  $0 \in D(G)$  is a matrix-valued rational function. Let  $H_{[M;N]}$  denote the Hankel operator of a normalized strongly right-coprime factor  $[M; N]$ . Let  $\varepsilon \in (0, \sqrt{1 - \|H_{[M;N]}\|^2})$ . Then there exists a rational  $\varepsilon$ -robust right factor stabilizing admissible feedback function for  $G$ .*

*Proof.* This follows from the proof of Proposition 8.17, using that in the proof of Proposition 8.15 we can choose  $[\tilde{V}, -\tilde{U}]$  rational by Lemma A.28.  $\square$

## Notes

We refer to Vidyagagar [94], Zhou, Doyle and Glover [103], Doyle, Francis and Tannenbaum [22] and Francis [28] for general information on stabilizing feedback functions for rational functions.

For finite-dimensional input and output spaces Proposition 8.9 is originally due to Inouye [40]. An independent proof was given by Smith [88]. The general case, also valid for infinite-dimensional input and output spaces, was first proven by Mikkola [56], whose proof we followed. Proposition 8.10 is due to Youla, Jabr and Bongiorno [98] in the case of rational functions. Using controllers with internal loop this result in the general case is due to Curtain, Weiss and Weiss [17]. The results presented here on robust right factor stabilizing feedback functions are due to McFarlane and Glover [36], [54] in the rational case. Continuous-time versions of the general case are given in Curtain and Zwart [18, Section 9.4], Oostveen [64, Section 7.2] and Curtain [11]. All of these references closely follow the arguments in McFarlane and Glover as do we.